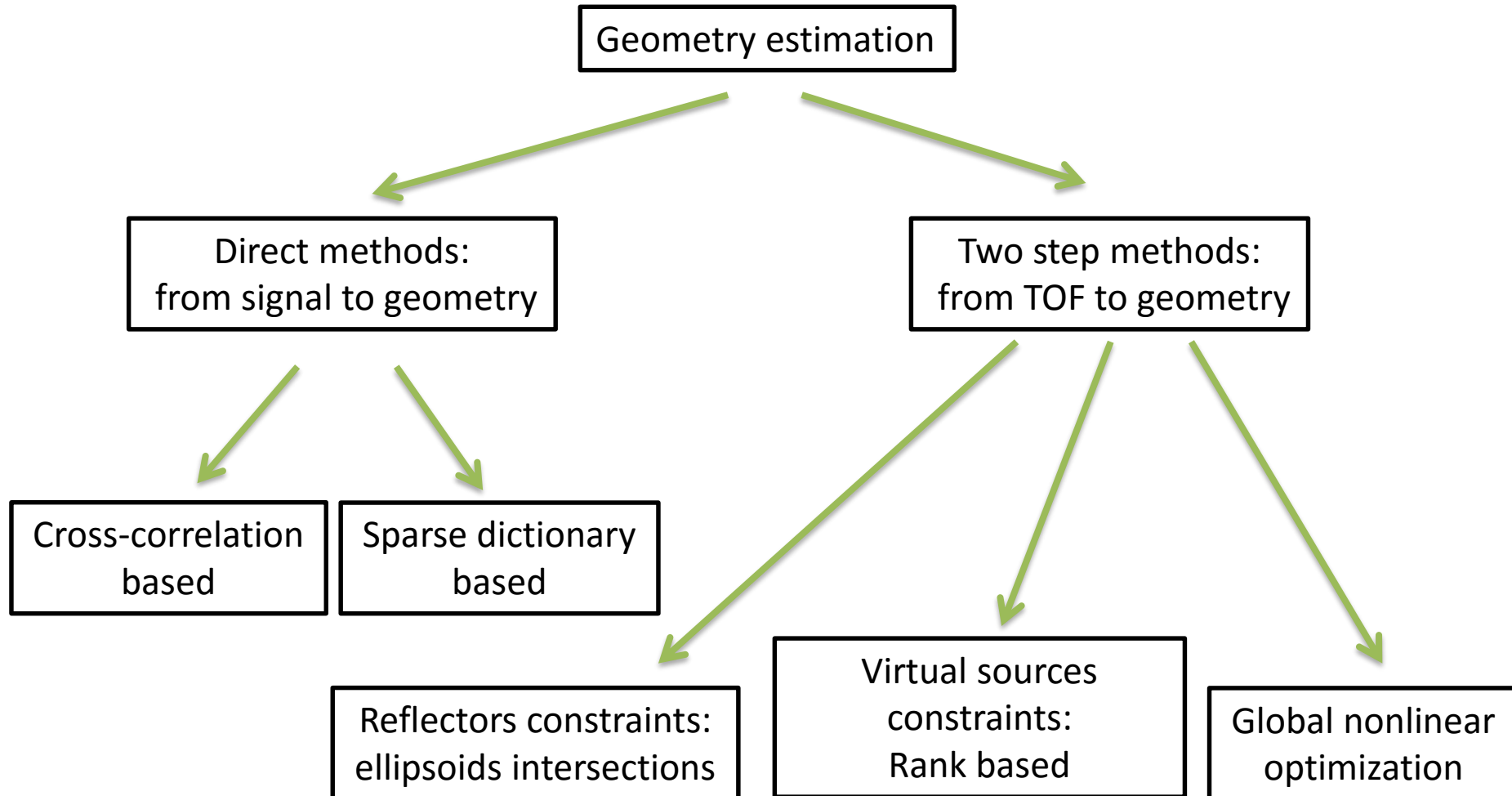
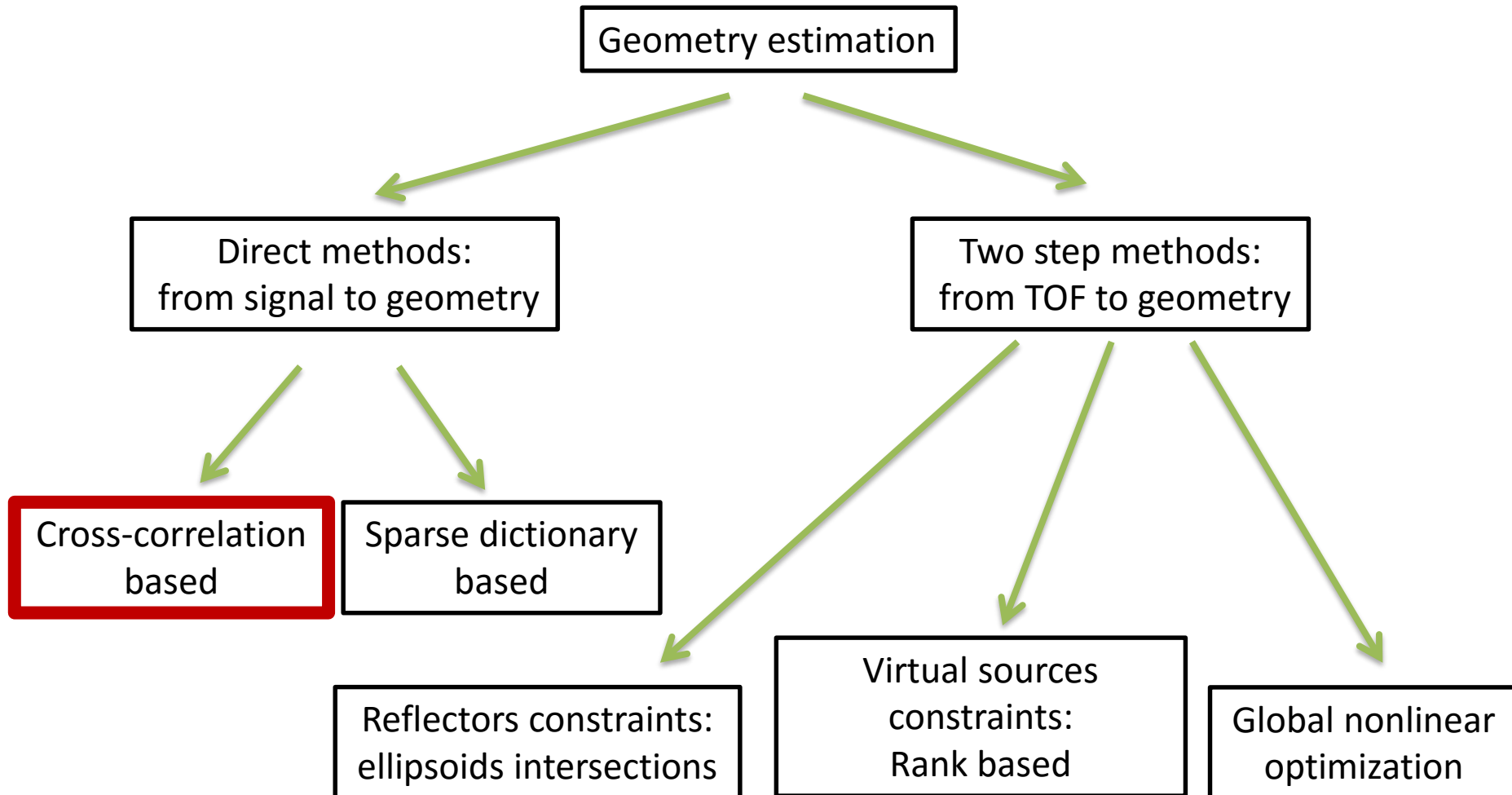


Room geometry estimation

Taxonomy

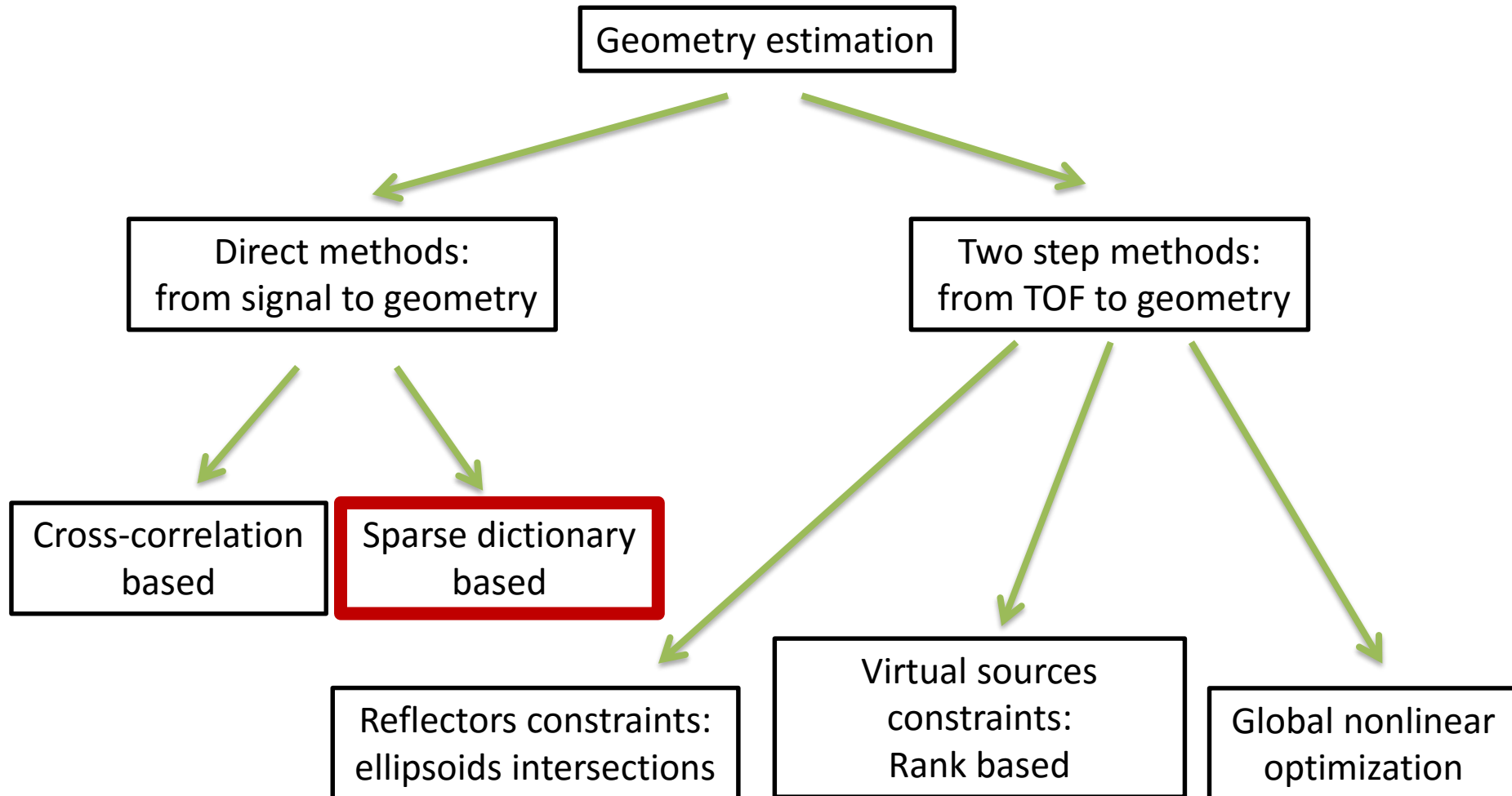


Taxonomy



Tervo & Korhonen, "Estimation of reflective surfaces from continuous signals." ICASSP 2010 .

Taxonomy

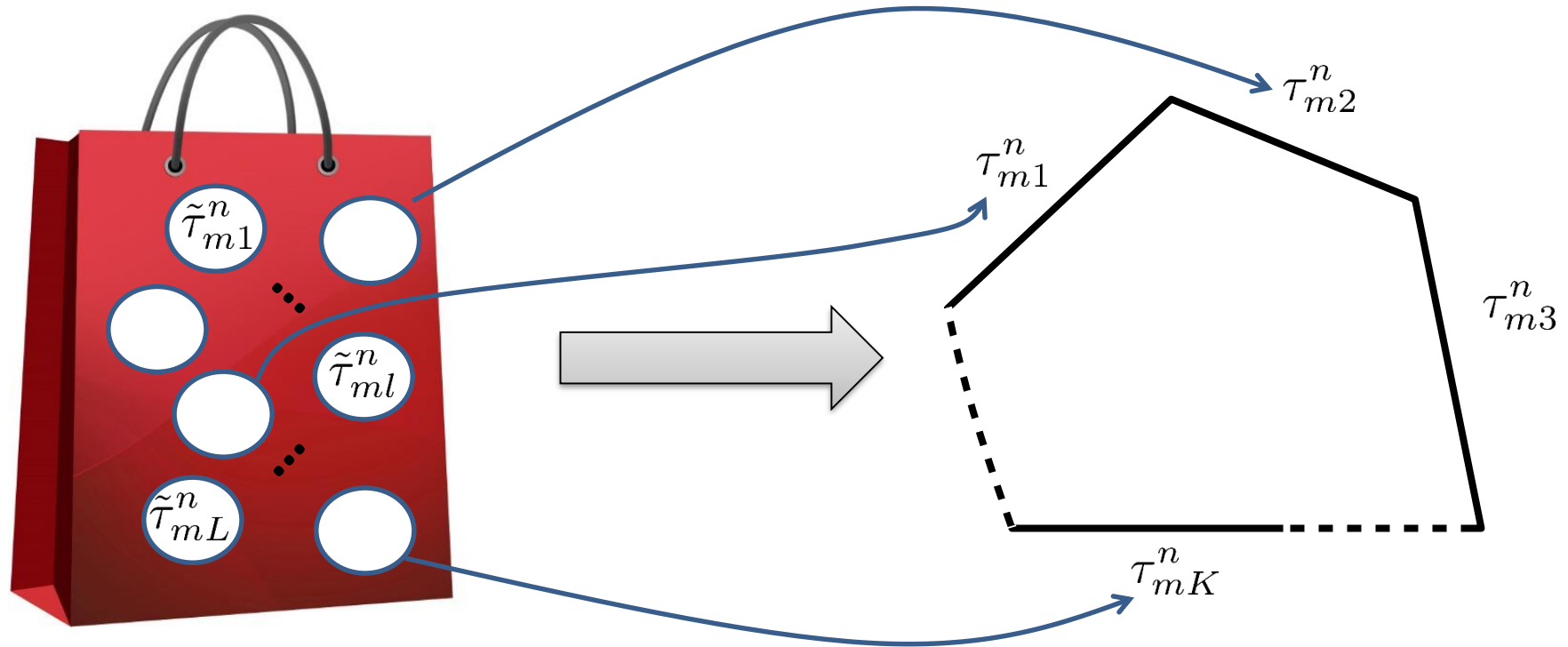


**Ribeiro et al., "Geometrically Constrained Room Modeling With Compact Microphone Arrays,"
IEEE Transactions on Audio, Speech, and Language Processing, 2012.**

Challenges

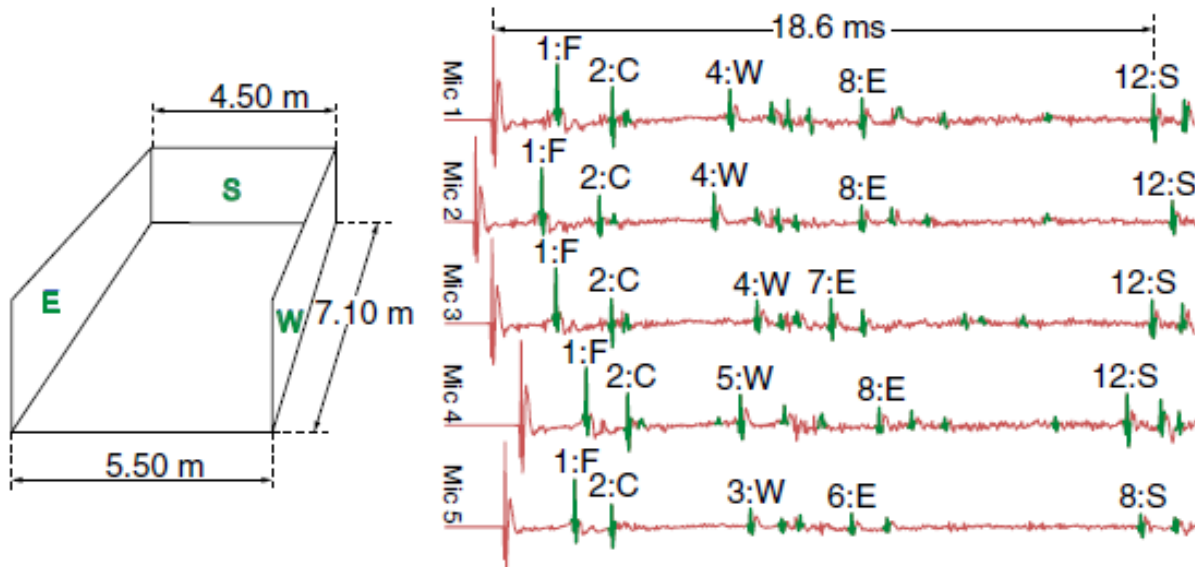
- Noise on measured delays;
- Spurious delays due to reflections from objects, diffraction effects, etc.;
- Missing delays due to low SNR measurements;
- **Unlabeled delays:** matching between planar reflectors and delays at each microphone is in general unknown.

Unlabeled delays



- Matching between delays and walls (echo sorting) has to be estimated.
- NP-hard permutation problem

Disambiguation of delays: a real case

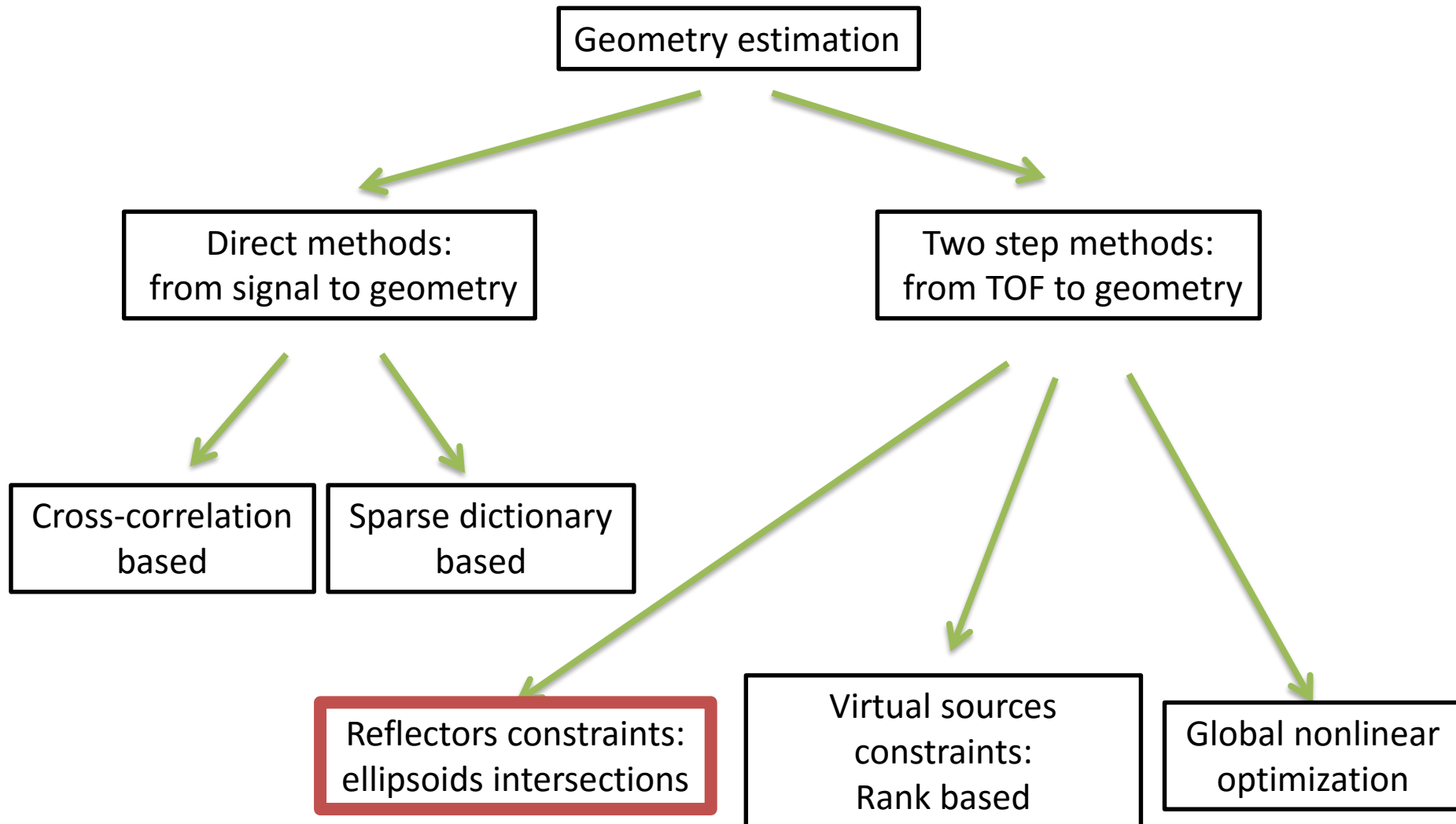


Courtesy of [Dokmanic et al. 2013]

After detecting delays in the real signals, the echo sorting problem assigns delays to ceiling (C), floor (F) and walls (E, S, W) of the room.

Second order delays may complicate further the problem with very spread configurations of microphones.

Taxonomy



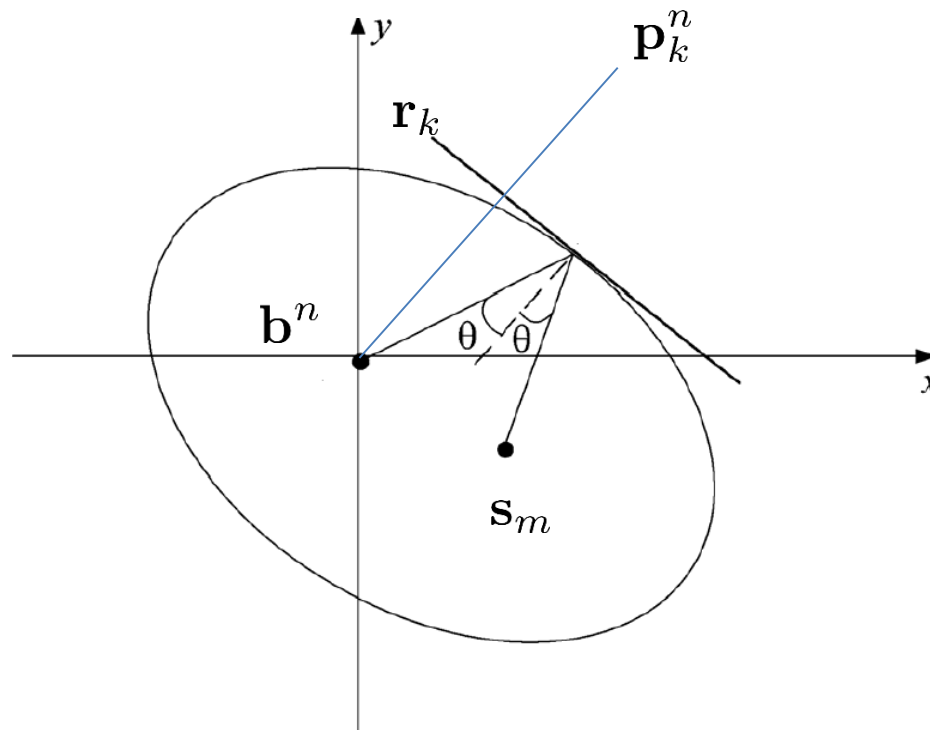
Reflectors constraints: ellipsoids intersections

Requirements:

- Sources and Microphones positions known;
- Tx signal known;
- Each source is placed very close to a planar reflector.

Reflector constraint (2D)

Given a TOF τ_{mk}^n extracted from a signal from microphone s_m and a source b^n , the corresponding reflector k is tangent to an ellipse of major diameter $d_{mk}^n = c\tau_{mk}^n$ and foci equal to b^n and s_m .



$$C(d_{mk}^n, b^n, s_m) = \{ \mathbf{x} : \|\mathbf{x} - b^n\|_2^2 + \|\mathbf{x} - s_m\|_2^2 = (d_{mk}^n)^2 \}$$

Conics in dual forms

Move to homogeneous coordinates: $\bar{\mathbf{x}}^\top = [\mathbf{x} \ 1]^\top$

Conics (ellipses) in homogeneous coordinates can be expressed as a 3 x 3 symmetric matrix:

$$\mathbf{C}_{mk}^n = \begin{bmatrix} a & b & c \\ b & d & e \\ c & e & f \end{bmatrix}$$

such that: $C(d_{mk}^n, \mathbf{s}_m, \mathbf{b}^n) = \bar{\mathbf{x}}^\top \mathbf{C}_{mk}^n \bar{\mathbf{x}}$

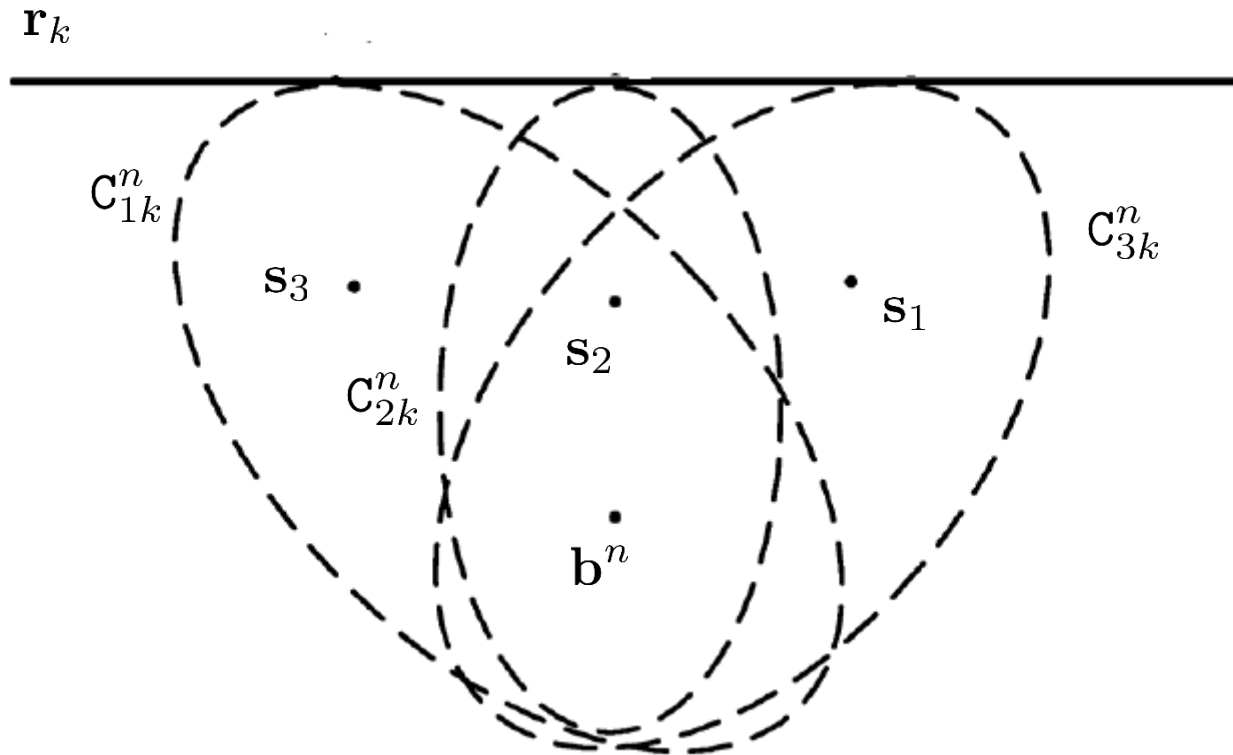
Dual form of a conic is parametrised by tangent lines to the conics rather than points.
This gives the following matrix form:

$$\tilde{\mathbf{C}}_{mk}^n = \mathbf{C}_{mk}^n^{-1}$$

with a constraint on the reflector line such that: $\bar{\mathbf{r}}_k^\top \tilde{\mathbf{C}}_{mk}^n \bar{\mathbf{r}}_k = 0$

where $\bar{\mathbf{r}}_k$ is the homogeneous form of the vector normal to the line, whose modulus is equal to the line distance from the origin: $\bar{\mathbf{r}}_k = [\mathbf{r}_k^\top \ 1]^\top$

Multiple microphones



Multiple ellipses from delays related to different microphones define a unique common tangent reflector. This requires the solution of the labelling problem!

Cost function of quadrics

In real cases the relation:

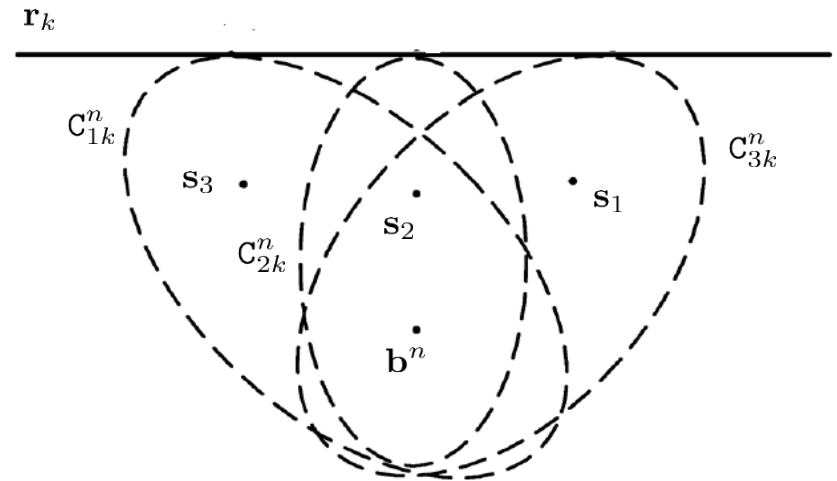
$$\bar{\mathbf{r}}_k^\top \tilde{\mathbf{C}}_{mk}^n \bar{\mathbf{r}}_k = 0$$

holds only approximately



Minimize the nonlinear cost function:

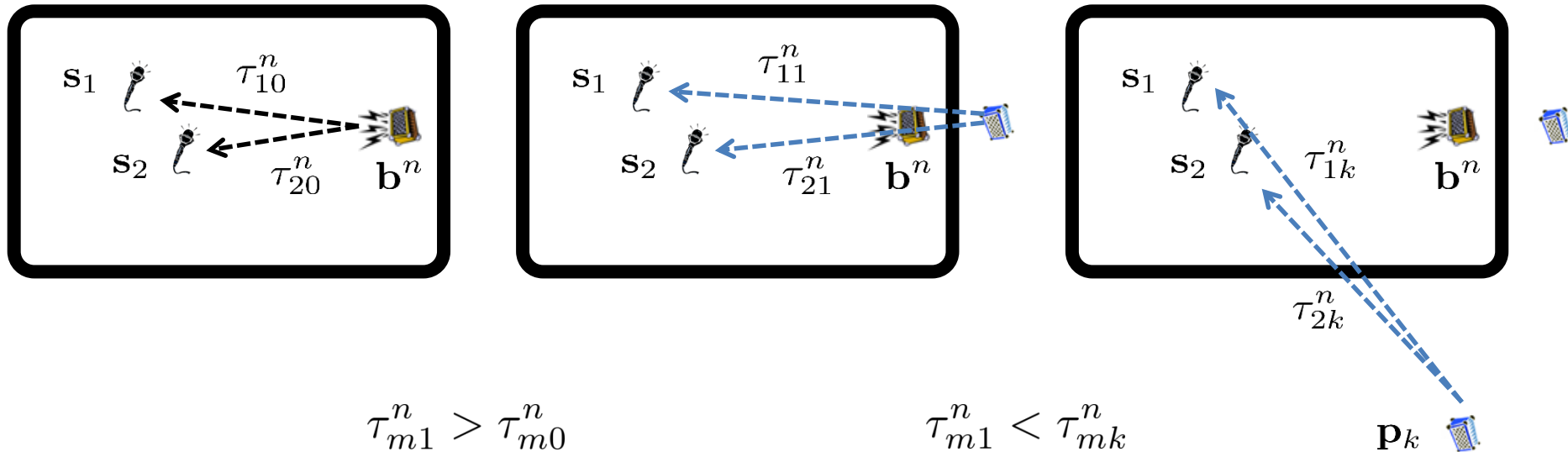
$$H(\bar{\mathbf{r}}) = \sum_{m=1}^M \|\bar{\mathbf{r}}^\top \tilde{\mathbf{C}}_{mk}^n \bar{\mathbf{r}}^\top\|_2^2$$



- One of the homogeneous coordinates fixed in order to avoid the trivial solution.
- Taking the gradient = 0, one obtains a fourth order polynomial equations systems with a finite number of solutions.
- Keep the solution giving the lowest value of the cost function.

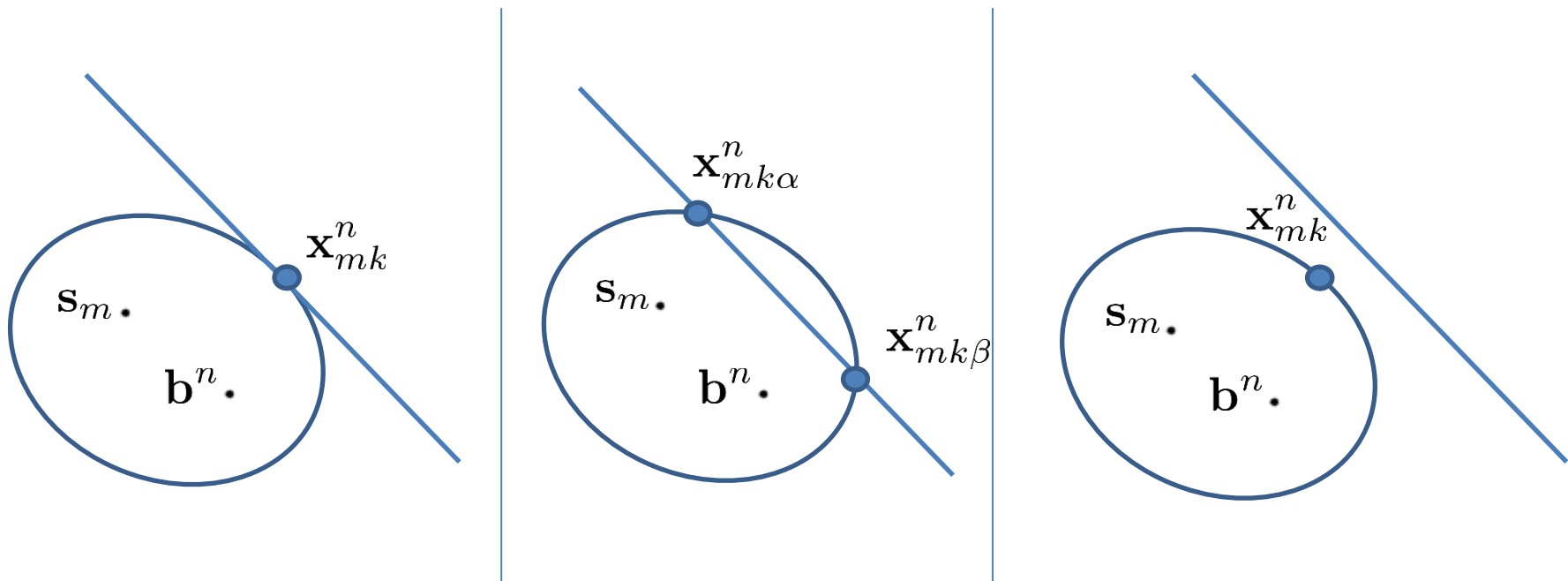
Labeling problem: proposed solution

- For each reflector put the source as close as possible to it. In this way the first delay after the direct path should belong to the same reflector for all the microphones.
- Estimate the line reflector from the collected set of delays from the above procedure.
- Repeat the procedure for each reflector.



Solution Refinement (1)

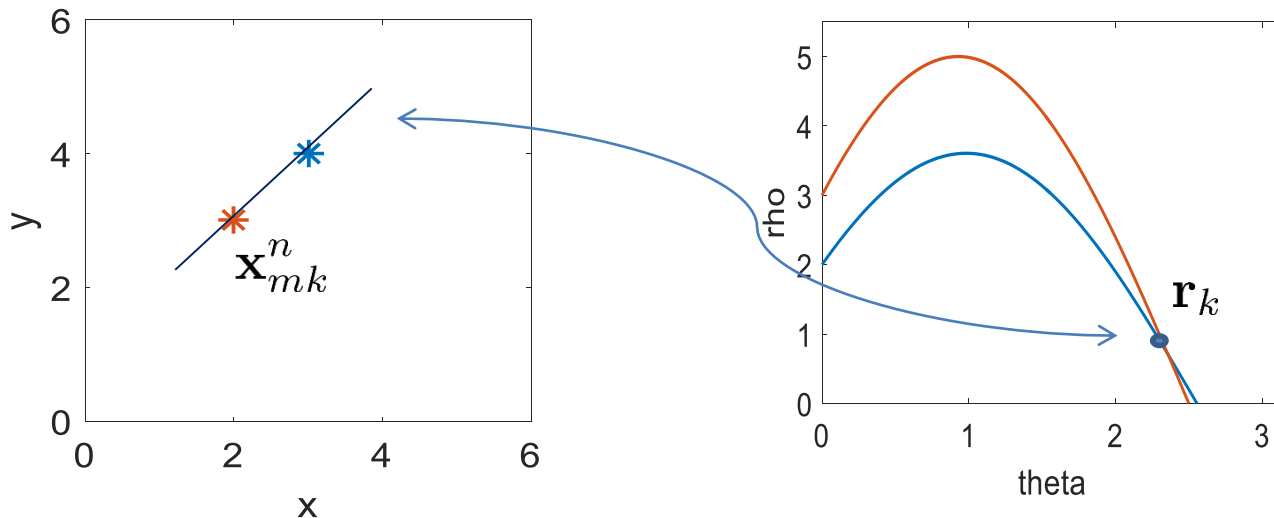
Given an estimated reflector k , evaluate for each ellipse nmk the intersecting, tangent, or most close point to the reflector.



Refinement procedure (2): Hough Transform

Hough transform maps points on curves in (ρ, θ) space. If a set of points belongs to the same line, the corresponding curves will intersect in a single point.

$$\rho = x \cos(\theta) + y \sin(\theta)$$

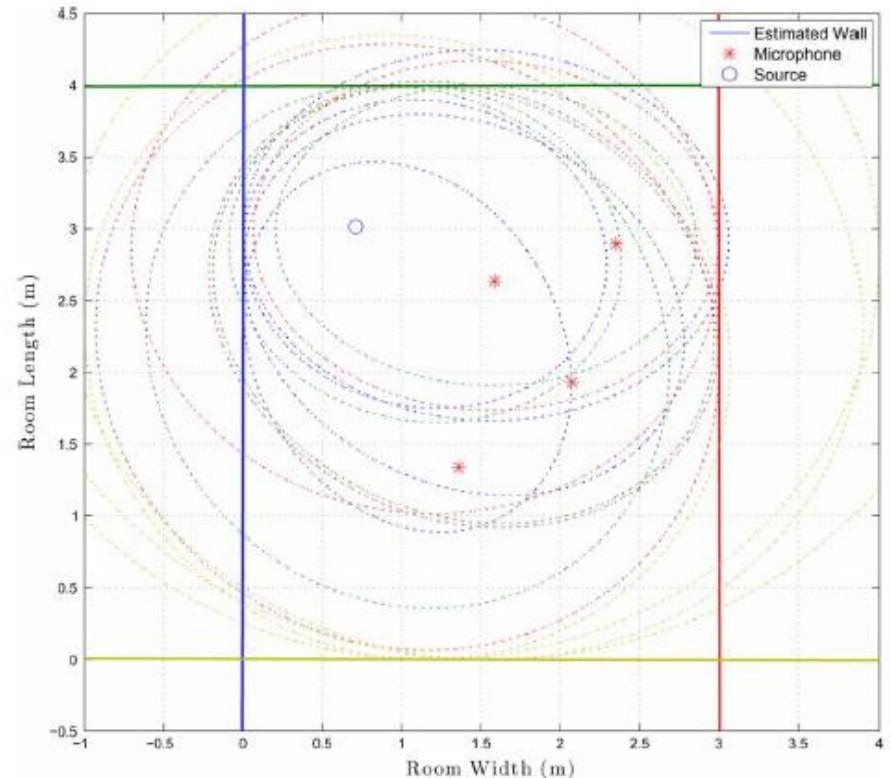


Refinement procedure (2)

Collect all the points \mathbf{x}_{mk}^n and perform Hough transform discretizing the curves into a grid in (theta, rho) space.

Increment a counter for each grid point crossed by a curve in the transformed space.

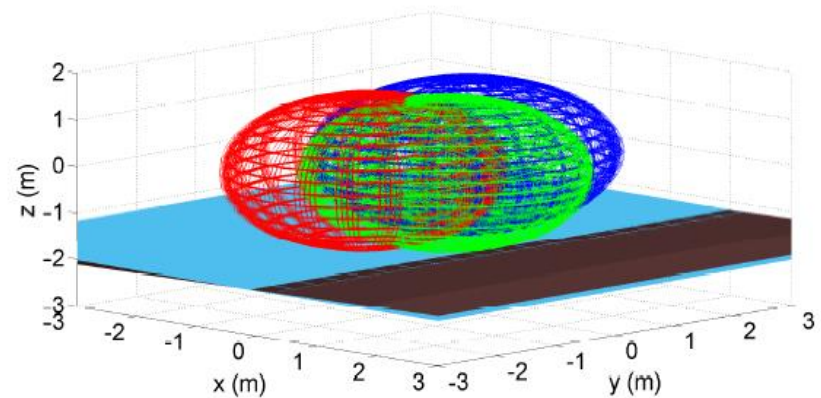
Take the position of the K largest maxima in (theta, rho) space as the refined positions for the K reflectors.



Extension to 3D: ellipsoids

In 3D space ellipses becomes ellipsoids, and linear reflectors become planar reflectors

$$Q_{mk}^n = \begin{bmatrix} a & b & c & d \\ b & e & f & g \\ c & f & h & i \\ d & g & i & l \end{bmatrix}$$



$$\tilde{Q}_{mk}^n = Q_{mk}^n{}^{-1}$$

$$\bar{\mathbf{r}}_k^\top \tilde{Q}_{mk}^n \bar{\mathbf{r}}_k = 0$$

$$H(\bar{\mathbf{r}}) = \sum_{m=1}^M \|\bar{\mathbf{r}}^\top \tilde{Q}_{mk}^n \bar{\mathbf{r}}^\top\|_2^2$$

Pros and Cons

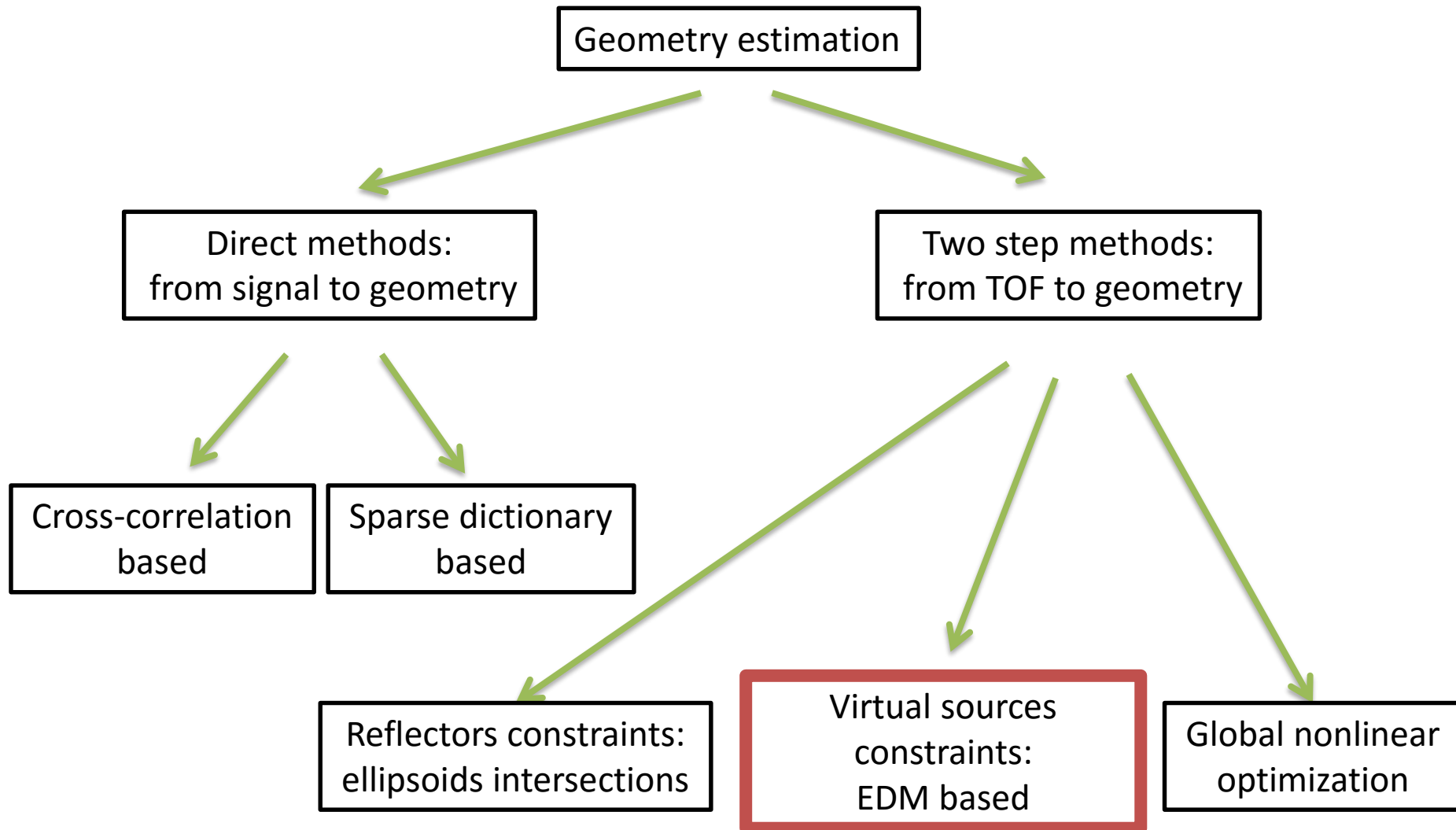
Pros

- Robustness to outliers (spurious delays) and echo labeling problem solved by local maxima search in the Hough space.

Cons

- Requires knowledge of both microphone and source positions.
- Source must be placed very close to each planar reflector.

Taxonomy



Virtual sources constraints: EDM based

Requirements:

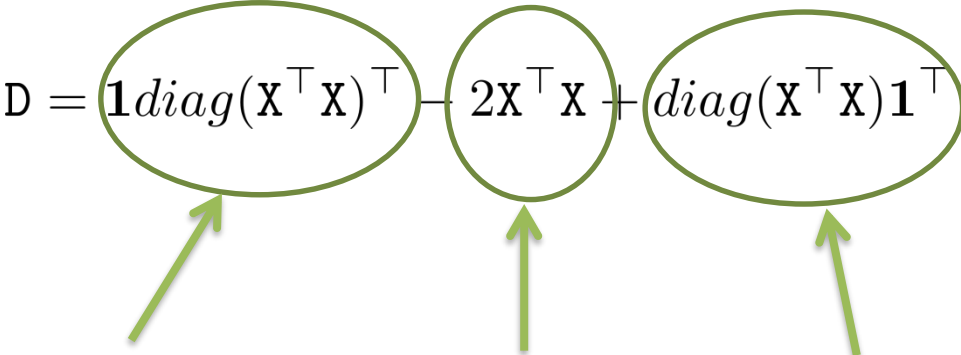
- One source, at least 4 microphones;
- Relative position among microphones assumed to be known.

Euclidean Distance Matrices EDM

Given a set of 3D points $\mathbf{x}_i, i = 1, \dots, I$ an Euclidean Distance Matrix EDM is defined as a matrix D of pairwise squared Euclidean distances between points:

$$D[i, j] = \|\mathbf{x}_i - \mathbf{x}_j\|_2^2$$

EDM has rank at most equal to 5:

$$D = \underbrace{\mathbf{1} \text{diag}(\mathbf{X}^\top \mathbf{X})^\top}_{\text{Rank} = 1} - \underbrace{2\mathbf{X}^\top \mathbf{X}}_{\text{Rank} = 3} + \underbrace{\text{diag}(\mathbf{X}^\top \mathbf{X})\mathbf{1}^\top}_{\text{Rank} = 1}$$


$$\mathbf{X} = [\mathbf{x}_1 \quad \dots \quad \mathbf{x}_I] \in \mathcal{R}^{I \times 3}$$

$$\mathbf{1} \in \mathcal{R}^{I \times 1}$$

Augmented EDM

Build the EDM from the pairwise (known) microphone-microphone distances

$$D[m_1, m_2] = \|\mathbf{s}_{m_1} - \mathbf{s}_{m_2}\|_2^2$$

Now, suppose to know the delay labeling and build a vector of distances related to the same reflector k :

$$\mathbf{d}_k = [d_{1k}^2, d_{2k}^2, \dots, d_{Mk}^2]$$

Build an augmented matrix $D_{aug}(\mathbf{d}_k)$ with the distances between the microphones and the k -virtual source .

$$D_{aug} = \begin{bmatrix} D & \mathbf{d}_k \\ \mathbf{d}_k^\top & 0 \end{bmatrix}$$

In absence of measurement errors on distances the matrix $\text{rank} = 5$ if and only if all the distances in vector \mathbf{d}_k belong to the same virtual source k .

Rank based approach (1)

Build a distance vector picking one TOF for each microphone:

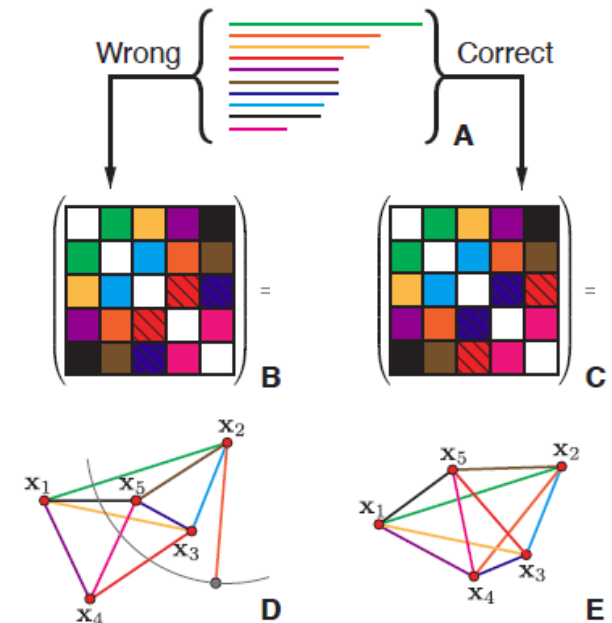
$$\mathbf{d}_{(i_1, \dots, i_M)} = [d_{1i_1}^2, d_{2i_2}^2, \dots, d_{Mi_M}^2]$$

d_{mi_m} : distance related to the TOF i_m from microphone m

Build an augmented distance matrix:

$$D_{aug}(i_1, \dots, i_M) = \begin{bmatrix} \mathbf{D} & \mathbf{d}_{(i_1, \dots, i_M)} \\ \mathbf{d}_{(i_1, \dots, i_M)}^\top & 0 \end{bmatrix}$$

Check the matrix rank of $D_{aug}(i_1, \dots, i_M)$ and retain the index set giving $\text{rank } D_{aug}(i_1, \dots, i_M) < 6$



If just 4 microphones are available, a modified augmented EDM with rank < 5 can be exploited, by subtracting a common row or column from D_{aug} .

S-stress approach

Rank checking is not robust to errors and noise in the TOF estimation



rely on S-stress measure:

$$s(i_1, \dots, i_M) = \min_{\mathbf{E}} \|\mathbf{D}(i_1, \dots, i_M) - \mathbf{E}\|_2^2, s.t. \mathbf{E} \in \mathbb{EDM}$$

$$\mathbf{E}[i, j] = \|\mathbf{x}_i - \mathbf{x}_j\|_2^2$$

S-stress measure of how close the matrix of measured squared distances is to an EDM

The method ranks in ascending order the s-stress cost given by the above problem and it keeps the index sets yielding the best results.

Testing for all the index sets may be cumbersome, need to rely on heuristics:

Distance between TOFs from different microphones related to the same planar cannot be higher than the maximum distance between pairs of microphones.

Solve for virtual sources

For each selected index set, pick the corresponding distances and solve for the virtual source positions

$$d_{mk}^2 = \|\mathbf{s}_m - \mathbf{p}_k\|_2^2 = \|\mathbf{s}_m\|_2^2 - 2\mathbf{s}_m^\top \mathbf{p}_k + \|\mathbf{p}_k\|_2^2$$

$$\tilde{d}_{mk}^2 = -\frac{1}{2} (d_{mk}^2 - \|\mathbf{s}_m\|_2^2) = \mathbf{s}_m^\top \mathbf{p}_k - \frac{1}{2} \|\mathbf{p}_k\|_2^2$$

$$\begin{bmatrix} \tilde{d}_{1k}^2 \\ \tilde{d}_{2k}^2 \\ \vdots \\ \tilde{d}_{Mk}^2 \end{bmatrix} = \begin{bmatrix} \mathbf{s}_1^\top & -\frac{1}{2} \\ \mathbf{s}_2^\top & -\frac{1}{2} \\ \vdots & \\ \mathbf{s}_M^\top & -\frac{1}{2} \end{bmatrix} \begin{bmatrix} \mathbf{p}_k \\ \|\mathbf{p}_k\|_2^2 \end{bmatrix} \quad \Rightarrow \quad \tilde{\mathbf{d}}_k = \mathbf{S} \begin{bmatrix} \mathbf{p}_k \\ \|\mathbf{p}_k\|_2^2 \end{bmatrix}$$

$$\begin{bmatrix} \mathbf{p}_k \\ \|\mathbf{p}_k\|_2^2 \end{bmatrix} = \mathbf{S}^\dagger \tilde{\mathbf{d}}_k$$

Given the virtual sources it is straightforward to recover corresponding reflectors \mathbf{r}_k

Pros and Cons

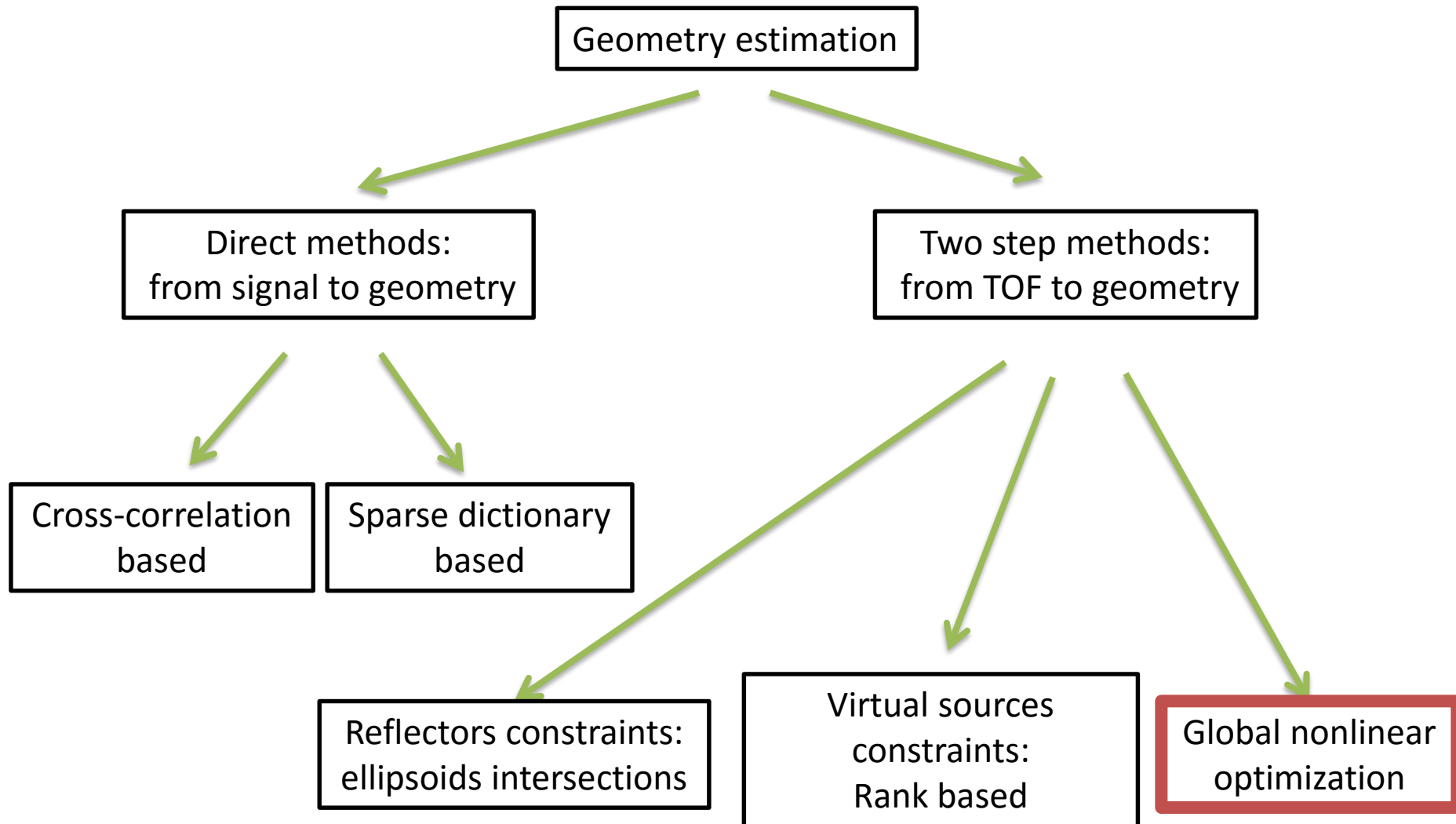
Pros

- Arbitrary (known) microphone displacements;
- Needs only first order reflections.

Cons

- Knowledge of relative microphone positions
- Evaluation of all TOF combinations may be cumbersome: heuristics needed to prune the selected sets.

Taxonomy



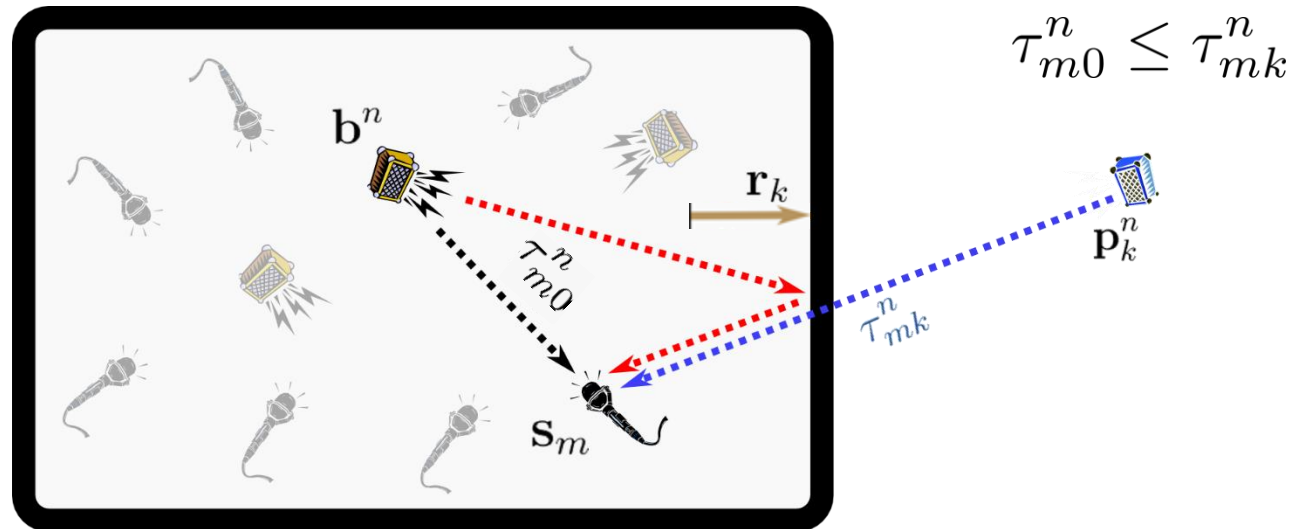
Global nonlinear optimization

Requirements:

- RX signals at multiple microphones from multiple sources.
- Number of planar reflectors known.
- **Not needed:** knowledge of microphone and sources position and emission and offset times (**fully uncalibrated method**).

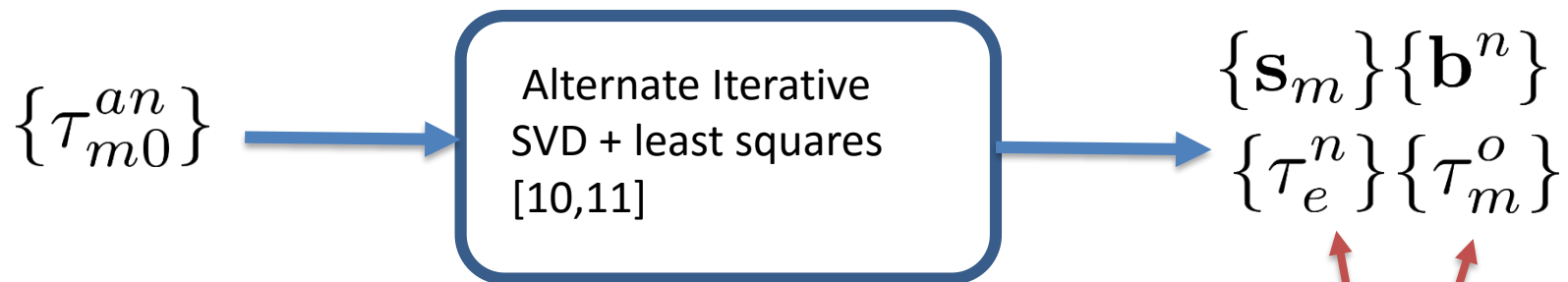
Exploit direct path properties

For each mic and source (m,n) the smallest delay of arrival corresponds to the direct path



Sources and mics positions from direct path delays

For each couple n, m sort the $K+1$ delays in ascending order and take the first one.



[11] M. Crocco, A. Del Bue, and V. Murino, "A bilinear approach to the position self-calibration of multiple sensors," *IEEE Transactions on Signal Processing*, vol. 60, pp. 660–673, 2012.

[10] Nikolay D. Gaubitch, W. Bastiaan Kleijn, and Richard Heusdens, "Auto-localization in ad-hoc microphone arrays," in *Acoustics, Speech and Signal Processing (ICASSP), 2013 IEEE International Conference on*, 2013, pp. 106–110.

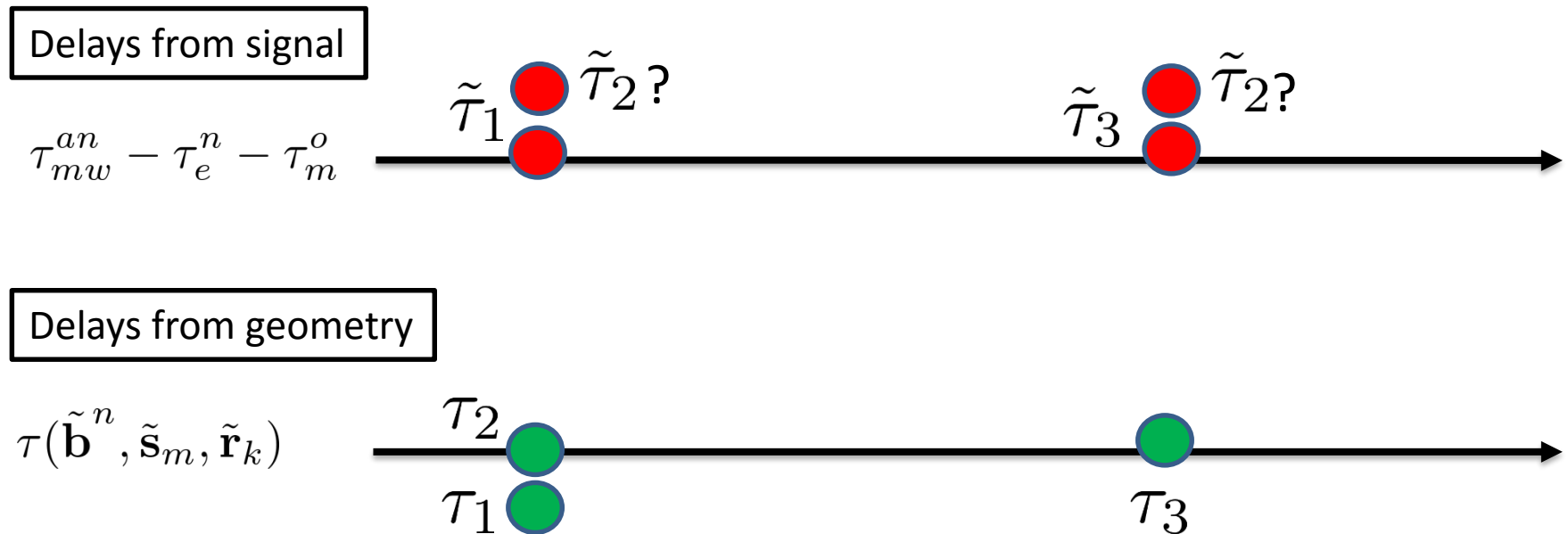
Bilinear factorization problem
(requires knowledge of emission and offset times)

Grounded on [11] : estimates also signal emission and offset times

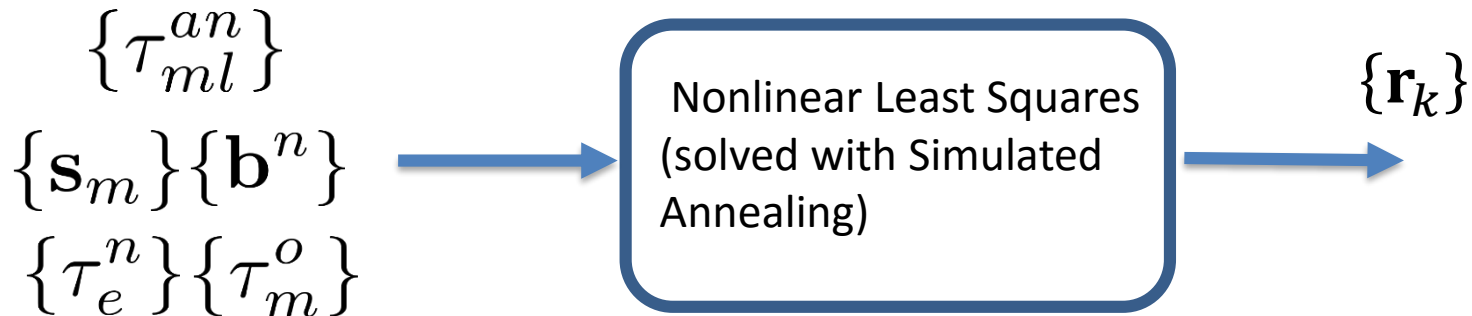
Ill-conditioning in presence of close delays

- If two or more delays are close to each other the problem is ill-conditioned.
- Same cost function minimum for a wrong delay reconstruction

Example of wrong delay reconstruction:



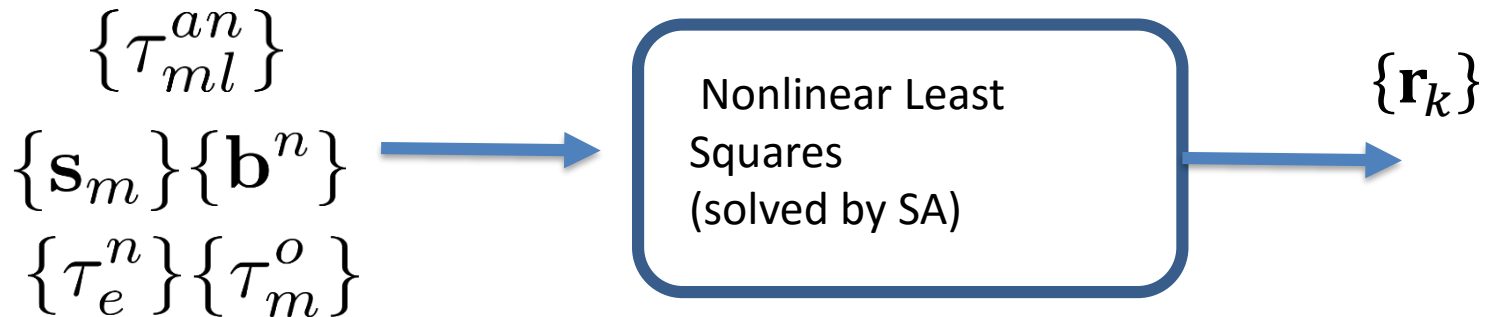
First guess estimation of walls coordinates



$$\min_{\mathbf{r}_k} \sum_{nm} I(n, m) \sum_k \left(\tau_{mk}^{an} - \tau_e^n - \tau_m^o - \tau(\tilde{\mathbf{b}}^n, \tilde{\mathbf{s}}_m, \mathbf{r}_{h(k)}) \right)$$

Index function sorting the set of delays in ascending order.
Matching problem between walls and delays is bypassed.

First guess estimation of walls coordinates

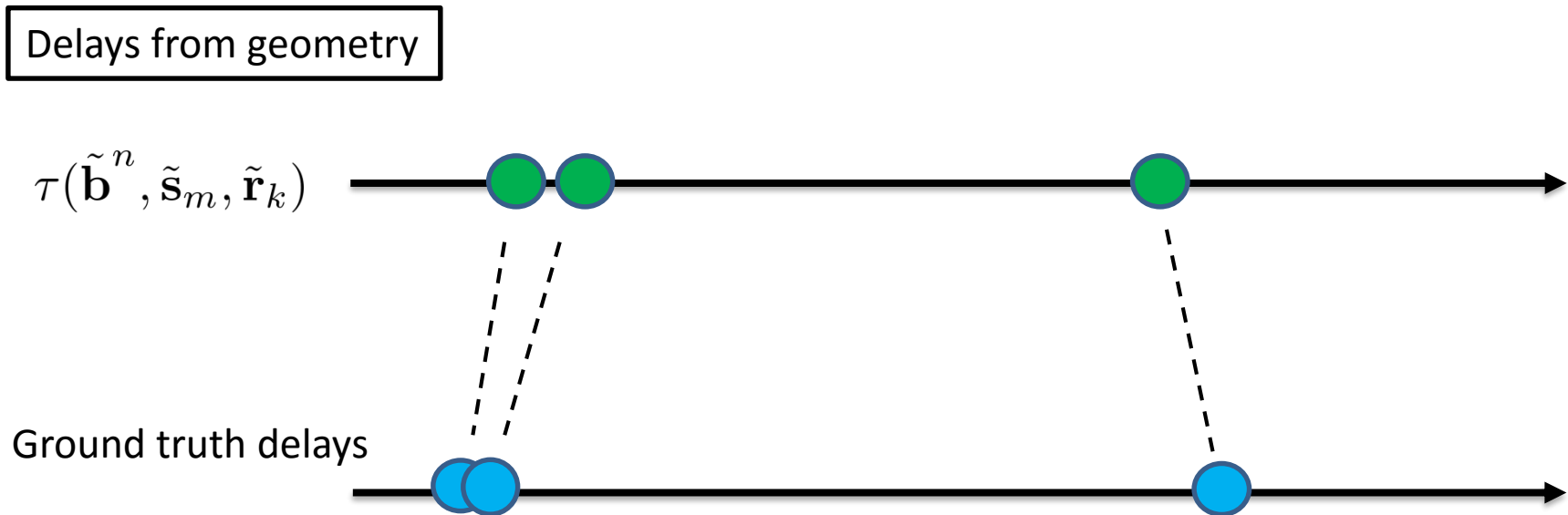


$$\min_{\mathbf{r}_k} \sum_{nm} I(n, m) \sum_k \left(\tau_{mk}^{an} - \tau_e^n - \tau_m^o - \tau(\tilde{\mathbf{b}}^n, \tilde{\mathbf{s}}_m, \mathbf{r}_{h(k)}) \right)$$

Pruning strategy: if $\tau(\tilde{\mathbf{b}}^n, \tilde{\mathbf{s}}_m, \mathbf{r}_{h(k_1)}) \approx \tau(\tilde{\mathbf{b}}^n, \tilde{\mathbf{s}}_m, \mathbf{r}_{h(k_2)})$ for some k_1, k_2 : $I(n, m) = 0$.
Terms containing possible wrong delays estimations are pruned out.

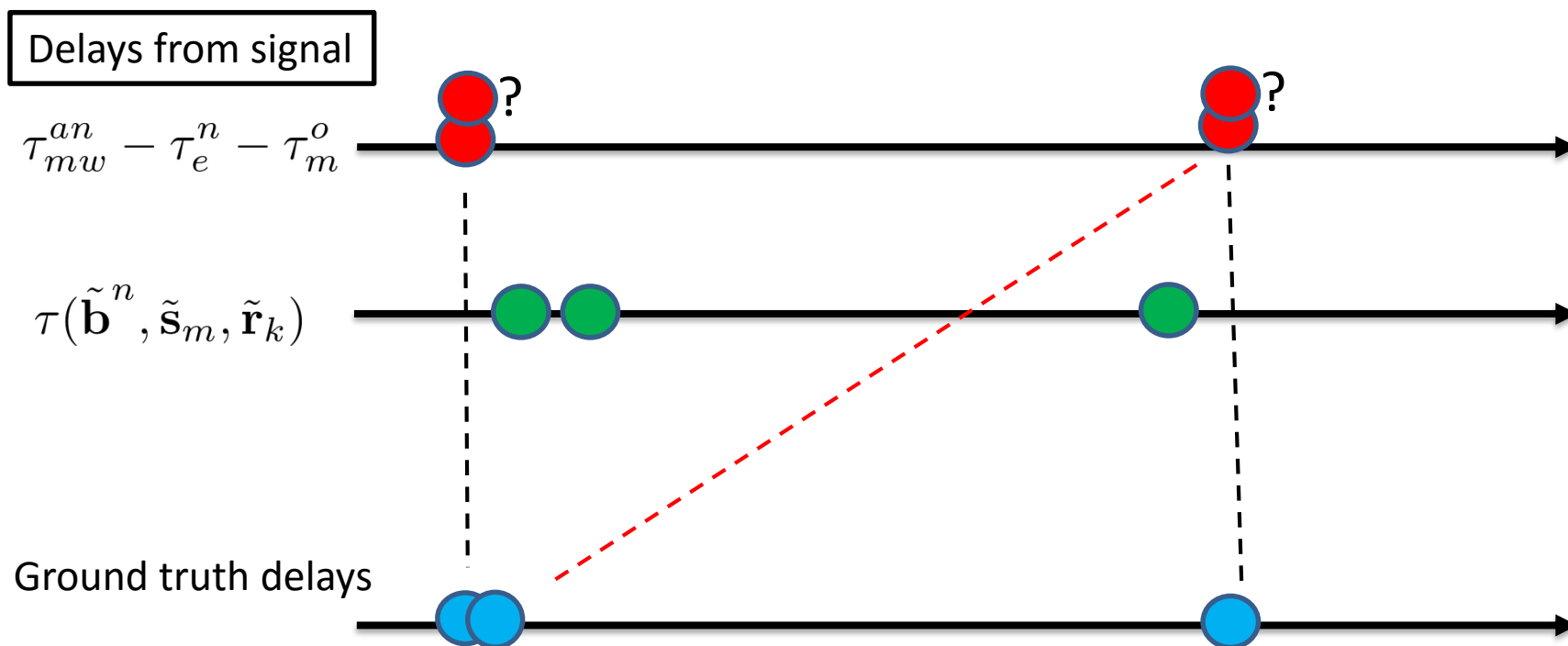
Fetching back ambiguous delays (1)

Delays estimated from geometry are not precise due to error accumulation along the procedure but are not subject to ambiguities.



Fetching back ambiguous delays (2)

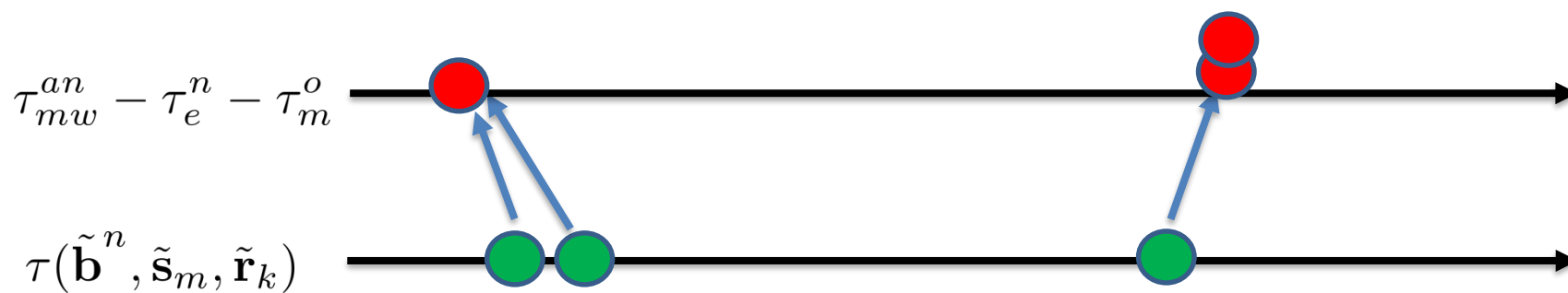
Delays estimated from the signals are more precise but are subject to ambiguities.



Fetching back ambiguous delays (3)

Nearest neighbour procedure

$$\tilde{w}_k = \arg \min_w \left(\tau_{mw}^{an} - \tau_e^n - \tau_m^o - \tau(\tilde{\mathbf{b}}^n, \tilde{\mathbf{s}}_m, \tilde{\mathbf{r}}_k) \right)$$



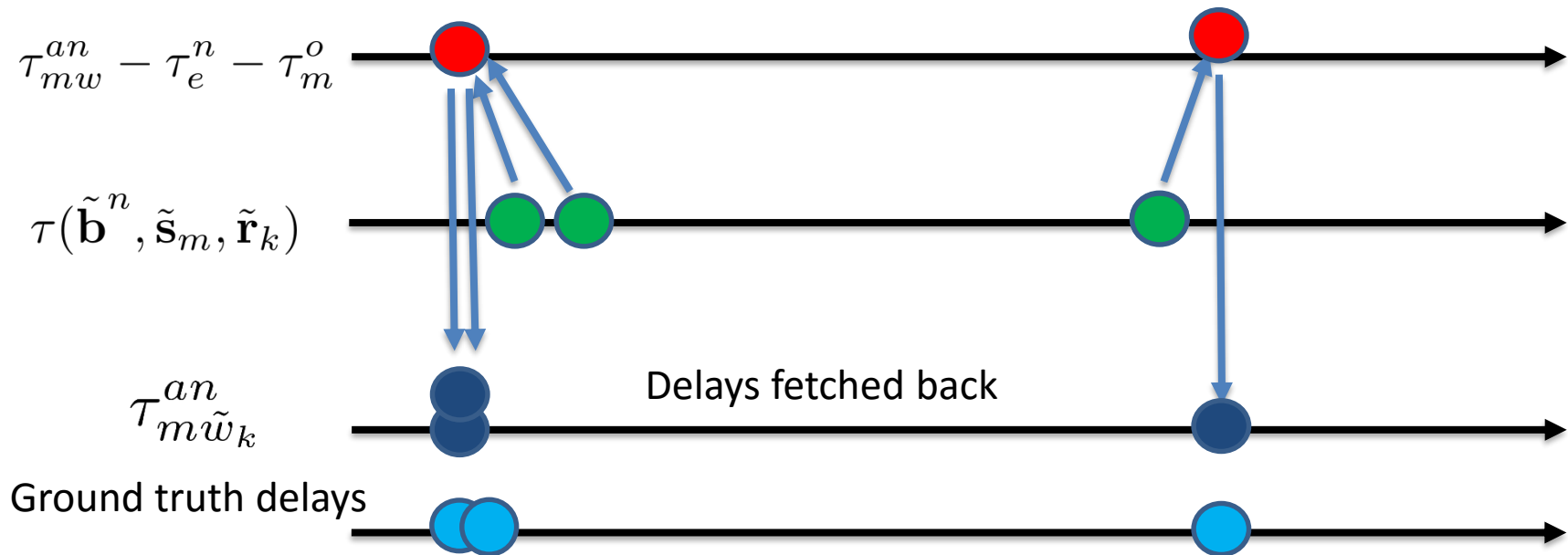
Ground truth delays



Fetching back ambiguous delays (4)

$$\tilde{w}_k = \arg \min_w \left(\tau_{mw}^{an} - \tau_e^n - \tau_m^o - \tau(\tilde{\mathbf{b}}^n, \tilde{\mathbf{s}}_m, \tilde{\mathbf{r}}_k) \right)$$

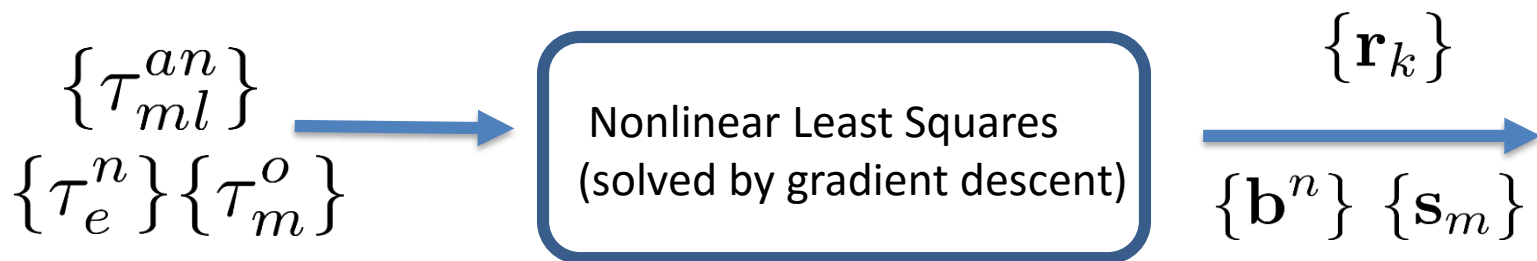
Final delays recovered back are precise and with no ambiguities



Final geometric optimization

Jointly estimate **walls, sources and microphones** positions by using all the set of delays

$$\min_{\mathbf{b}^n, \mathbf{s}_m, \mathbf{r}_k} \sum_{nmk} \left(\tau_{m\tilde{w}_k}^{an} - \tau_e^n - \tau_m^o - \tau(\mathbf{b}^n, \mathbf{s}_m, \mathbf{r}_k) \right)^2$$



Pros and Cons

Pros

- Fully uncalibrated method: no required knowledge of microphone and sources positions, as well as TX emission and RX offset times.
- Delay labeling problem bypassed by sorting delays at each iteration of simulated annealing.

Cons

- Nonlinear, non-convex cost function involved, no guarantee to find the global maximum.
- Computationally demanding due to simulated annealing procedure.
- Limited robustness to missing or spurious delays (the problem might be alleviated adopting different loss functions like Huber).

NEXT: Dataset & evaluation